LF411 Low Offset, Low Drift JFET Input Operational Amplifier



Literature Number: SNOSBH6C

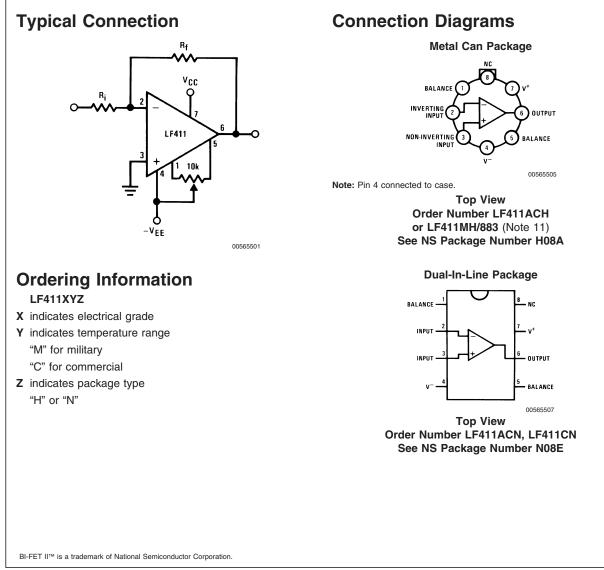


LF411 Low Offset, Low Drift JFET Input Operational Amplifier General Description Features

These devices are low cost, high speed, JFET input operational amplifiers with very low input offset voltage and guaranteed input offset voltage drift. They require low supply current yet maintain a large gain bandwidth product and fast slew rate. In addition, well matched high voltage JFET input devices provide very low input bias and offset currents. The LF411 is pin compatible with the standard LM741 allowing designers to immediately upgrade the overall performance of existing designs.

These amplifiers may be used in applications such as high speed integrators, fast D/A converters, sample and hold circuits and many other circuits requiring low input offset voltage and drift, low input bias current, high input impedance, high slew rate and wide bandwidth.

Internally trimmed offset voltage:	0.5 mV(max)
Input offset voltage drift:	10 µV/°C(max)
Low input bias current:	50 pA
Low input noise current:	0.01 pA/√Hz
Wide gain bandwidth:	3 MHz(min)
High slew rate:	10V/µs(min)
Low supply current:	1.8 mA
High input impedance:	10 ¹² Ω
Low total harmonic distortion:	≤0.02%
Low 1/f noise corner:	50 Hz
Fast settling time to 0.01%:	2 µs



August 2000

Absolute Maximum Ratings (Note 1)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/ Distributors for availability and specifications.

	LF411/	A LF411
Supply Voltage	±22V	±18V
Differential Input Volta	age ±38V	±30V
Input Voltage Range		
(Note 2)	±19V	±15V
Output Short Circuit		
Duration	Continuo	us Continuous
	H Package	N Package
Power Dissipation		
(Notes 3, 10)	670 mW	670 mW

	H Package	N Package			
T _j max	150°C	115°C			
θ _j Α	162°C/W (Still Air)	120°C/W			
	65°C/W (400				
	LF/min				
	Air Flow)				
θ _j C	20°C/W				
Operating Temp.					
Range	(Note 4)	(Note 4)			
Storage Temp.					
Range	–65°C≤T _A ≤150°C -	-65°C≤T _A ≤150°C			
Lead Temp.					
(Soldering,	260°C	260°C			
10 sec.)					
ESD Tolerance	erance Rating to be determined.				

DC Electrical Characteristics (Note 5)

Symbol	Parameter	Conditions		Parameter Conditions LF411A				Units		
				Min	Тур	Max	Min	Тур	Мах	Ī
Vos	Input Offset Voltage	R _s =10 kΩ, T _A =25	°C		0.3	0.5		0.8	2.0	mV
$\Delta V_{OS} / \Delta T$	Average TC of Input	R _s =10 kΩ (Note 6	6)		7	10		7	20	µV/°C
	Offset Voltage								(Note 6)	
I _{os}	Input Offset Current	V _s =±15V	Tj=25℃		25	100		25	100	рА
		(Notes 5, 7)	Tj=70°C			2			2	nA
			T _j =125°C			25			25	nA
I _B	Input Bias Current	V _s =±15V	T _i =25℃		50	200		50	200	рА
		(Notes 5, 7)	T _i =70°C			4			4	nA
			T _i =125°C			50			50	nA
R _{IN}	Input Resistance	T _j =25°C	·		10 ¹²			10 ¹²		Ω
A _{VOL}	Large Signal Voltage	V _S =±15V, V _O =±1	0V,	50	200		25	200		V/mV
	Gain	R _L =2k, T _A =25°C								
		Over Temperature	9	25	200		15	200		V/mV
Vo	Output Voltage Swing	V _S =±15V, R _L =10	K	±12	±13.5		±12	±13.5		V
V _{CM}	Input Common-Mode			±16	+19.5		±11	+14.5		V
	Voltage Range				-16.5			-11.5		V
CMRR	Common-Mode	R _s ≤10k		80	100		70	100		dB
	Rejection Ratio									
PSRR	Supply Voltage	(Note 8)		80	100		70	100		dB
	Rejection Ratio									
Is	Supply Current				1.8	2.8		1.8	3.4	mA

AC Electrical Characteristic (Note 5)

Symbol	Parameter	Conditions	LF411A		LF411			Units	
			Min	Тур	Max	Min	Тур	Max	
SR	Slew Rate	V _S =±15V, T _A =25°C	10	15		8	15		V/µs
GBW	Gain-Bandwidth Product	V _S =±15V, T _A =25°C	3	4		2.7	4		MHz
		T _A =25°C, R _S =100Ω, f=1 kHz		25			25		nV/√Hz
in Equivalent Input Noise Current		T _A =25°C, f=1 kHz		0.01			0.01		pA/√Hz

AC Electrical Characteristic (Note 5) (Continued)

Symbol	Parameter	Conditions	LF411A		LF411			Units	
		Min	Тур	Мах	Min	Тур	Мах		
THD	Total Harmonic Distortion	n $A_V = +10, R_L = 10k,$		<0.02			<0.02		%
		V _O =20 Vp-p,							
		BW=20 Hz-20 kHz							

Note 1: "Absolute Maximum Ratings" indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is functional, but do not guarantee specific performance limits.

Note 2: Unless otherwise specified the absolute maximum negative input voltage is equal to the negative power supply voltage.

Note 3: For operating at elevated temperature, these devices must be derated based on a thermal resistance of $\theta_i A$.

Note 4: These devices are available in both the commercial temperature range $0^{\circ}C \leq T_A \leq 70^{\circ}C$ and the military temperature range $-55^{\circ}C \leq T_A \leq 125^{\circ}C$. The temperature range is designated by the position just before the package type in the device number. A "C" indicates the commercial temperature range and an "M" indicates the military temperature range. The military temperature range is available in "H" package only.

Note 5: Unless otherwise specified, the specifications apply over the full temperature range and for $V_S=\pm 20V$ for the LF411A and for $V_S=\pm 15V$ for the LF411. V_{OS} , I_B , and I_{OS} are measured at $V_{CM}=0$.

Note 6: The LF411A is 100% tested to this specification. The LF411 is sample tested to insure at least 90% of the units meet this specification.

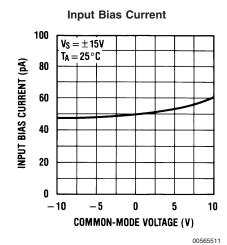
Note 7: The input bias currents are junction leakage currents which approximately double for every 10°C increase in the junction temperature, T_j . Due to limited production test time, the input bias currents measured are correlated to junction temperature. In normal operation the junction temperature rises above the ambient temperature as a result of internal power dissipation, P_D . $T_j=T_A+\theta_{jA}$ P_D where θ_{jA} is the thermal resistance from junction to ambient. Use of a heat sink is recommended if input bias current is to be kept to a minimum.

Note 8: Supply voltage rejection ratio is measured for both supply magnitudes increasing or decreasing simultaneously in accordance with common practice, from ±15V to ±5V for the LF411 and from ±20V to ±5V for the LF411A.

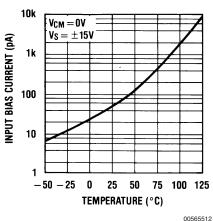
Note 9: RETS 411X for LF411MH and LF411MJ military specifications.

Note 10: Max. Power Dissipation is defined by the package characteristics. Operating the part near the Max. Power Dissipation may cause the part to operate outside guaranteed limits.

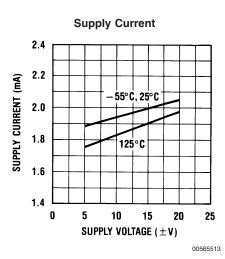
Typical Performance Characteristics



Input Bias Current

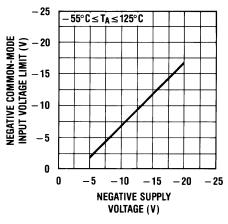


Typical Performance Characteristics (Continued)



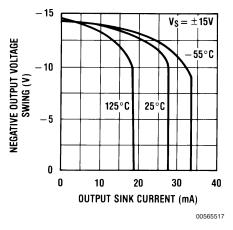


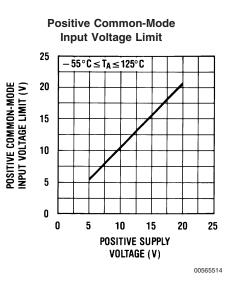




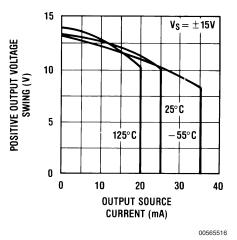




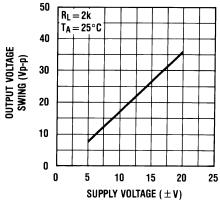




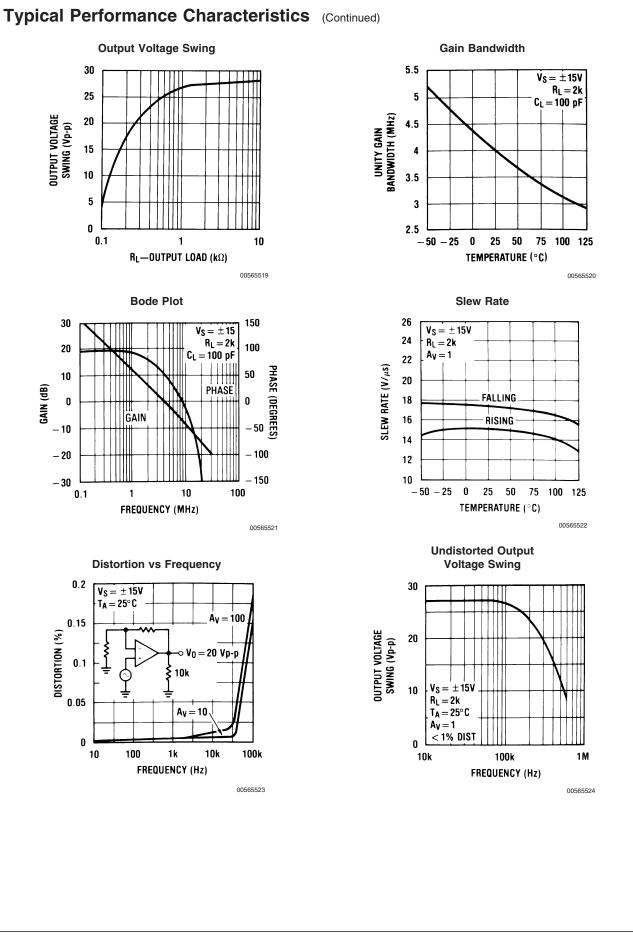
Positive Current Limit

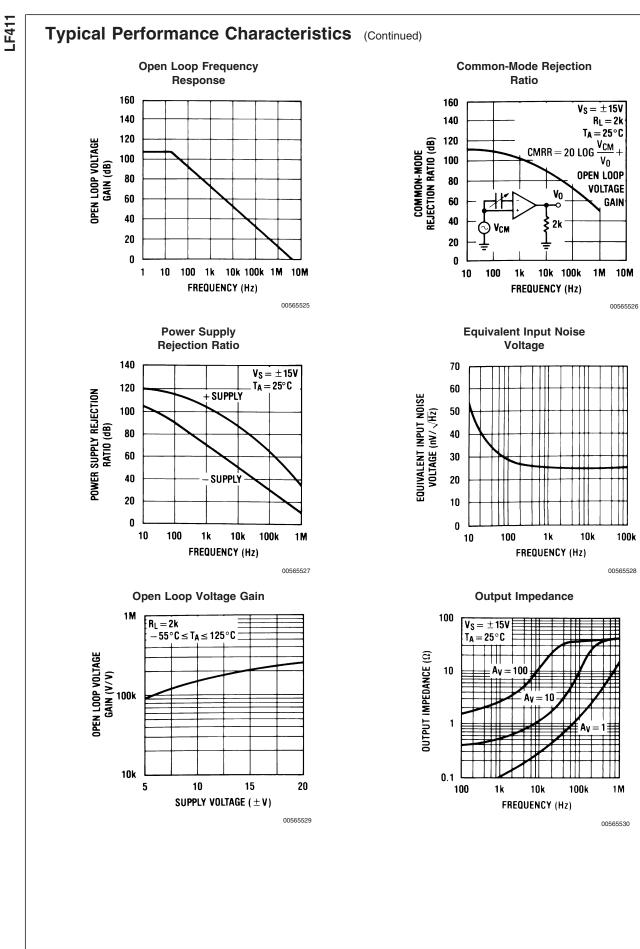


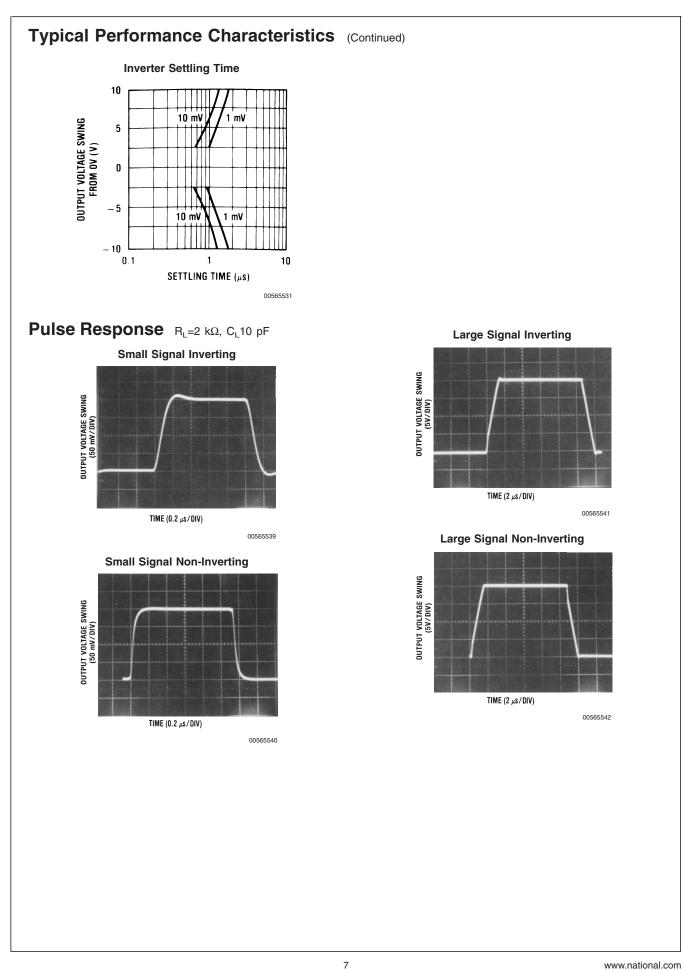
Output Voltage Swing



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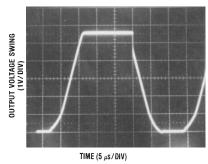






Pulse Response $R_L=2 k\Omega$, $C_L10 pF$ (Continued)

Current Limit (R_L=100Ω)



00565543

Application Hints

The LF411 series of internally trimmed JFET input op amps (BI-FET II™) provide very low input offset voltage and guaranteed input offset voltage drift. These JFETs have large reverse breakdown voltages from gate to source and drain eliminating the need for clamps across the inputs. Therefore, large differential input voltages can easily be accommodated without a large increase in input current. The maximum differential input voltage is independent of the supply voltages. However, neither of the input voltages should be allowed to exceed the negative supply as this will cause large currents to flow which can result in a destroyed unit.

Exceeding the negative common-mode limit on either input will force the output to a high state, potentially causing a reversal of phase to the output. Exceeding the negative common-mode limit on both inputs will force the amplifier output to a high state. In neither case does a latch occur since raising the input back within the common-mode range again puts the input stage and thus the amplifier in a normal operating mode.

Exceeding the positive common-mode limit on a single input will not change the phase of the output; however, if both inputs exceed the limit, the output of the amplifier may be forced to a high state.

The amplifier will operate with a common-mode input voltage equal to the positive supply; however, the gain bandwidth and slew rate may be decreased in this condition. When the negative common-mode voltage swings to within 3V of the negative supply, an increase in input offset voltage may occur.

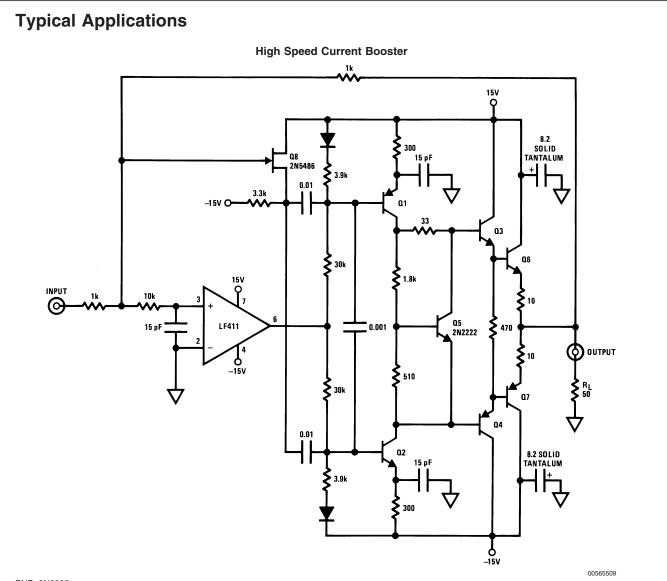
The LF411 is biased by a zener reference which allows normal circuit operation on $\pm 4.5V$ power supplies. Supply voltages less than these may result in lower gain bandwidth and slew rate.

The LF411 will drive a 2 k Ω load resistance to ±10V over the full temperature range. If the amplifier is forced to drive heavier load currents, however, an increase in input offset voltage may occur on the negative voltage swing and finally reach an active current limit on both positive and negative swings.

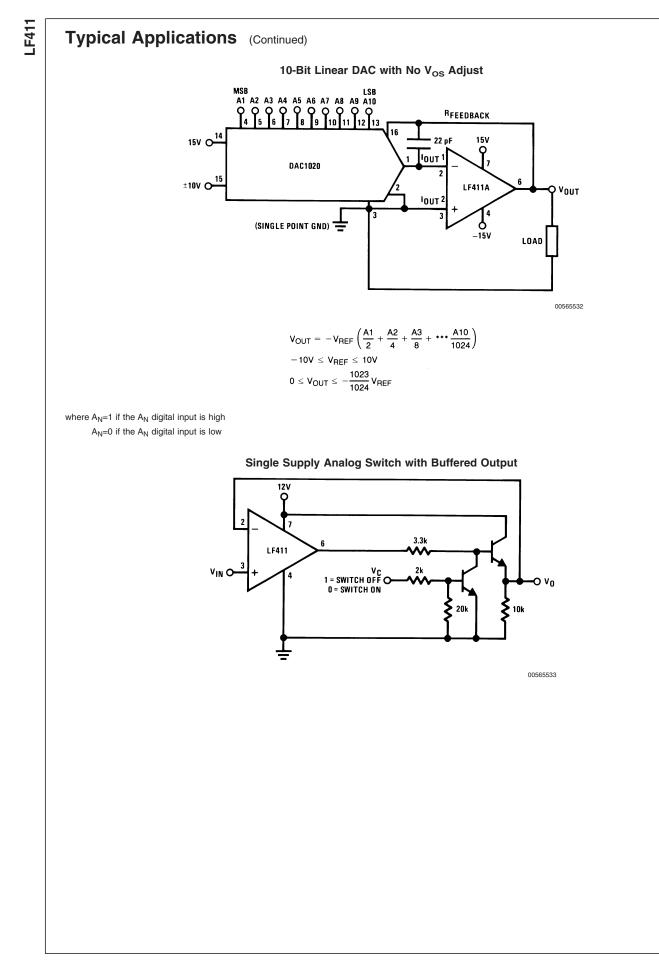
Precautions should be taken to ensure that the power supply for the integrated circuit never becomes reversed in polarity or that the unit is not inadvertently installed backwards in a socket as an unlimited current surge through the resulting forward diode within the IC could cause fusing of the internal conductors and result in a destroyed unit.

As with most amplifiers, care should be taken with lead dress, component placement and supply decoupling in order to ensure stability. For example, resistors from the output to an input should be placed with the body close to the input to minimize "pick-up" and maximize the frequency of the feedback pole by minimizing the capacitance from the input to ground.

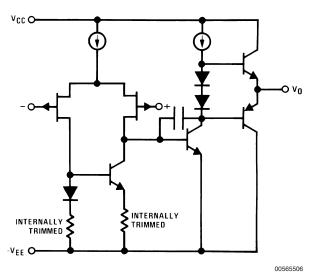
A feedback pole is created when the feedback around any amplifier is resistive. The parallel resistance and capacitance from the input of the device (usually the inverting input) to AC ground set the frequency of the pole. In many instances the frequency of this pole is much greater than the expected 3 dB frequency of the closed loop gain and consequently there is negligible effect on stability margin. However, if the feedback pole is less than approximately 6 times the expected 3 dB frequency, a lead capacitor should be placed from the output to the input of the op amp. The value of the added capacitor should be such that the RC time constant of this capacitor and the resistance it parallels is greater than or equal to the original feedback pole time constant.



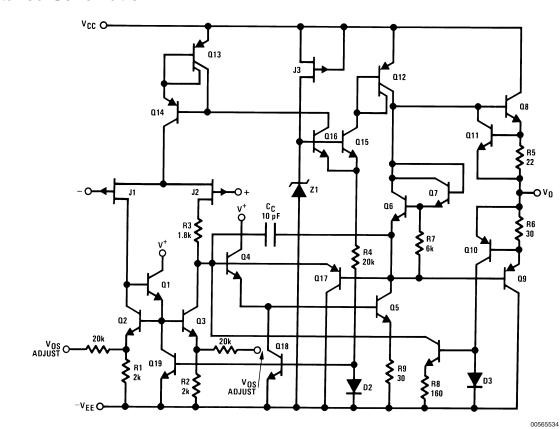
PNP=2N2905 NPN=2N2219 unless noted TO-5 heat sinks for Q6-Q7



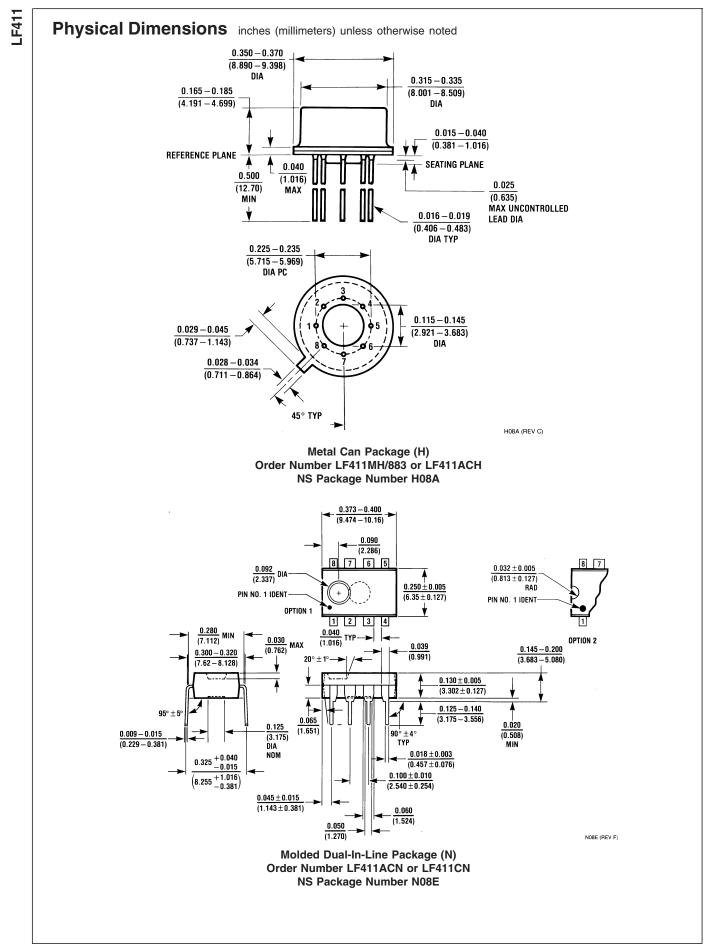
Simplified Schematic



Note 11: Available per JM38510/11904



Detailed Schematic



Notes

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